

Invited Paper

Experimental research on interaction of gyrotron traveling wave tube with inner mode converter in w-band TE₀₂ mode

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Abstract: This paper explicates the experimental research of W-band TE₀₂ mode gyrotron traveling wave tube interaction, explores the reasons for adopting TE₀₂ mode as the working mode, the existing competitive modes, the high-frequency structure adopted, the simulation of the interaction, and the design of the inner mode converter. Finally, the test results are given.

Keywords: W-band Gyro-TWT, TE₀₂ mode, Periodic dielectric loaded, Electron flow rate, Inner mode converter

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1. Introduction

In the millimeter wave band, compared with traveling wave tubes [1-5], the gyrotron traveling wave tube has the characteristics of high power and wide bandwidth. In the W-band, the output power of the traveling wave tube is generally less than 200W. The gyrotron traveling wave tube can generate output power of more than hundreds of kilowatts, with a bandwidth of more than 5%. Literature [5] reported the parameters of w-band gyrotron traveling wave tube: TE₀₁ mode, electron voltage 68kV, electron beam current 7A, magnetic field 3.4 Tesla, maximum peak output power of 135 kilowatts, 3dB bandwidth 4.1GHz. In the experiment, electron was found to bombard the high-frequency cut-off section. As the duty increased, the bombarding electrons seriously damaged the cut-off section structure. The copper bombarded by the electrons evaporated to the input window, causing the reflection of the input window to increase. The signal could not be injected, and the entire tube was damaged, as shown in Figure 1, Figure 2.

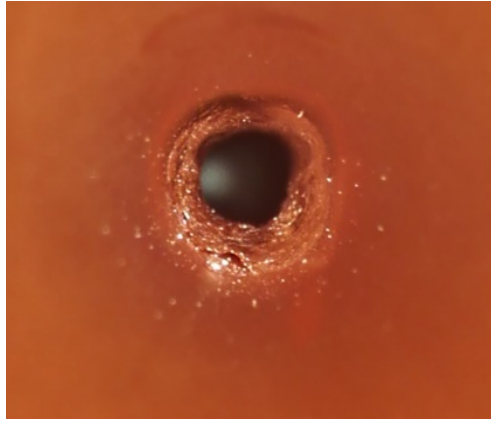


Fig. 1 Cut-off section damaged by electron bombardment



Fig. 2 The input window covered by the evaporated copper after electron bombardment

We have analyzed the reasons for electron bombardment to the high-frequency cut-off section:

1) In The W-band, gyrotron traveling wave tube frequency is high, and the interaction magnetic field in the uniform zone is required to reach $3.4T$, while the magnetic field in the cathode zone is about $1500-3000\text{ Gauss}$, which causes the length of adiabatic compression zone of the electron gun to be too long.

2) In The W-band, when gyrotron traveling wave tube is working in TE_{01} mode, the high-frequency interaction size is small, and the cut-off radius is smaller at this time.

3) In the experiment, when the Gyro-TWT tube was put into the superconducting magnet, it cannot be guaranteed that the axis of the initial tube body and the axis of the magnetic field are completely coaxial.

Due to the above three reasons, the Z axis along which the electrons were spiraling and the magnetic field axis are different, leading the electron beam to bombard the cut-off section, causing the evaporation of copper, therefore, polluting the window and the interaction section.

In order to improve the power capacity of the whole tube, the TE₀₂ mode is selected as the working mode of the W-band Gyro-TWT, which increases the size of the interaction, improves the flow rate of the electron beam, and avoids electron bombardment to the cut-off section. This paper analyzes the testing process.

2. Theoretical analysis

1) *Determination of structural parameters of high frequency interaction:*

Generally, the working voltage of the Gyro-TWT ranges from 60kV-70kV. We adopted 65kV and the working mode is TE₀₂. The operating point of weak relativity devices is generally selected near the cut-off frequency. Combining the working frequency and working mode, the waveguide radius is fixed at R=3.8cm, which is much larger than that of TE₀₁ mode.

In gyro-TWT, the amplification is achieved by using a cyclotron electron beam to interact with the transmitted microwave, which can be described by the following dispersion equations

$$k_z^2 = k^2 - k_c^2 \quad (1)$$

$$\omega = k_z + s\Omega \quad (2)$$

where S is the harmonic number, Ω is the angular frequency of the electron transverse circular motion, k_z , k and k_c are the axial wave number of the wave, the wave number in the free space and the cutoff wave number in the waveguide, respectively, and ω is the circular frequency of the wave. For general gyrotron devices, the magnitude of the applied DC magnetic field can be determined according to the above two synchronization conditions and working frequency:

$$B_0 \approx \frac{\omega m_0 \gamma}{es} \quad (3)$$

Get the working magnetic field $B_0=3.5T$.

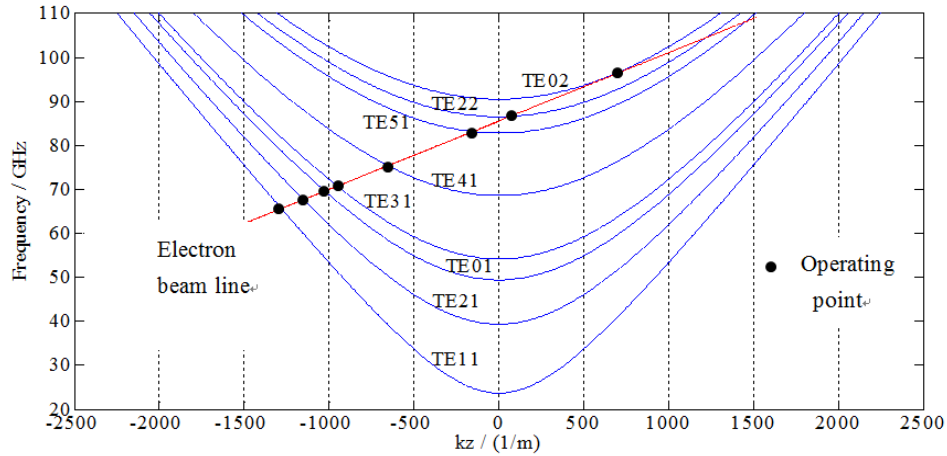


Fig. 3 Dispersion curve of TE₀₂ mode gyro-TWT

Basing on the above equation, the dispersion curve of TE₀₂ mode gyro amplifier is obtained. As shown in Figure3: Working mode TE₀₂, competitive mode E₂₂, TE₅₁, TE₄₁, TE₃₁, TE₀₁, TE₂₁ and TE₁₁, the frequencies are 86.5GHz, 83.3GHz, 76.2GHz, 72GHz, 70.9GHz, 69.2GHz and 67GHz, respectively. Therefore, how to suppress the competitive mode is the key to design TE₀₂ mode gyro amplifier.

2) Determination of guiding center radius

Fig.4 shows the relationship between the coupling coefficient of TE₀₂ mode and competition mode and the guiding center radius [6]. It can be seen from Figure 4 that for TE₀₂ mode, when the coupling coefficient is the largest, the guiding center radius is 0.26 times the interaction radius.

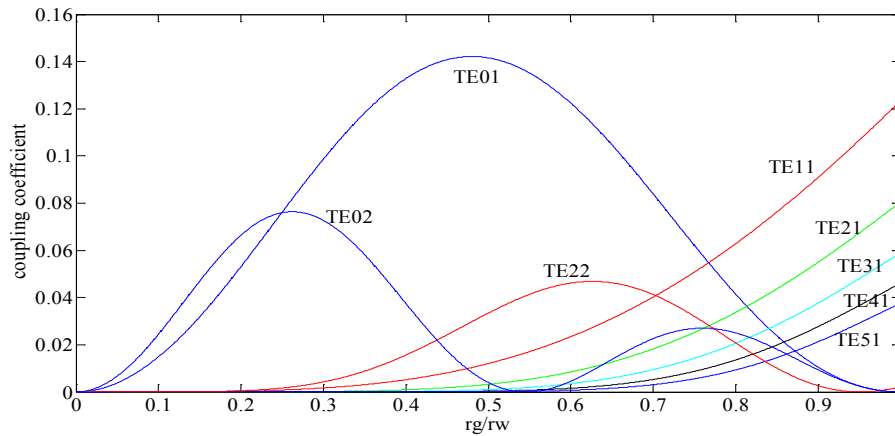


Fig. 4 The dependence of the coupling coefficient on guiding-center radius for TE₀₂ mode and possible competing modes

r_g is the guiding center radius, and r_w is the waveguide radius. It can also be seen from Fig. 4 that for TE_{02} mode, when the guiding center radius of the electron beam is $0.26 r_w$, the guiding center radius of other competition modes at the point of maximum coupling coefficient is far away from $0.26 r_w$, which helps suppress mode competition.

3) Determination of high frequency interaction structure:

The periodic dielectric loaded circular waveguide is used as the high frequency interaction system [7]. As shown in Figure 5, the conductor and the dielectric are interlaced and loaded by adjusting the proportion of the conductor and the dielectric and adjusting the radial thickness of the dielectric. This structure can well suppress the absolute instability and backward wave oscillation.

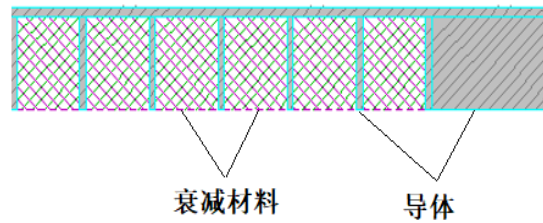


Fig. 5 Periodic medium loading structure.

3. Simulation results

Based on the above discussion, the data given in Table 1 are obtained:

Tab. 1 Interaction structure parameters

Waveguide radius	3.8mm	Relative permittivity of dielectric	13
Guide center radius	1mm	Loss tangent of dielectric	0.27
Magnetic field	3.4T	Medium length in cycle	15mm
Voltage	65Kv	Periodic conductor length	2mm
Electric current	10A	Conductor length	40mm

After optimized simulation, Fig. 6 shows that the electron beam produces stable pre-modulation when periodic medium loading, and the working mode enlarges when unloading. Fig. 7 shows the field distribution in the interaction system, from which it is clear that a stable TE_{02} mode is obtained in the amplification section of the interaction.

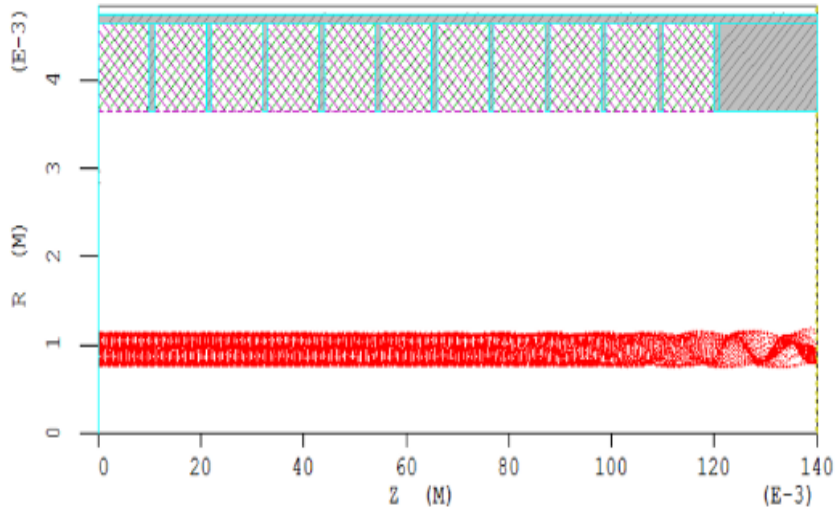


Fig.6 The electron beam trajectory obtained from simulation

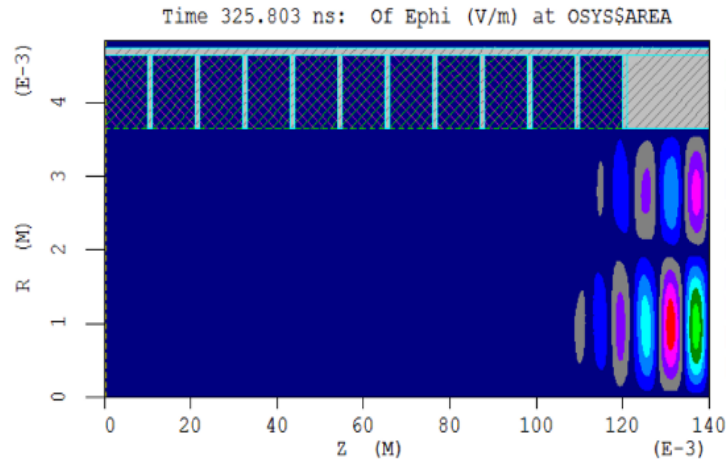


Fig. 7 The field distribution diagram obtained from simulation

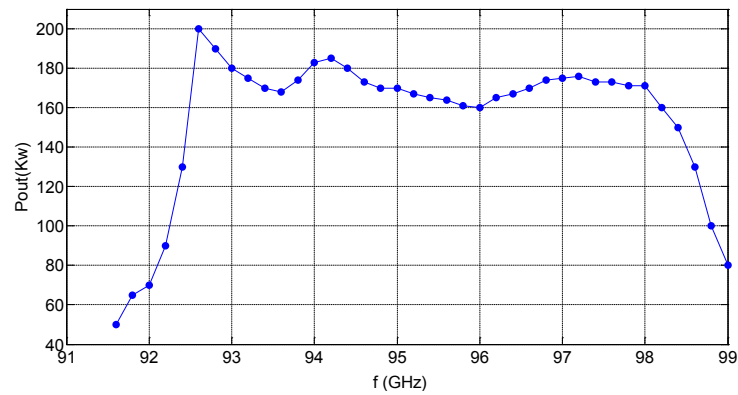


Fig. 8 Interaction power varying with frequency

Fig. 8 shows the dependence of the output power on the frequency for the W-band TE_{02} mode gyro-TWT. When the frequency is $92.6GHz$, the output power reaches $200kW$ and the $3dB$ bandwidth $6GHz$.

4. Experiment results

1) The design of the mode converter

Through calculation and experimental verification, after the beam-wave interaction, the electron beam falls into the collector region, and the corresponding magnetic field ranges from $2000 Gauss$ to $1000 Gauss$, as shown in the Figure 9. The distance of this area is $130mm$, and the corresponding distance from the port of the gradual output section to $2000 Gauss$ is $308mm$. Therefore, this space can be used to place the $TE_{02}-TE_{01}$ mode converter at the position of the collector to convert the working mode of the Gyro-TWT from TE_{02} mode to TE_{01} mode. In this way, the Gyro-TWT can output the TE_{01} mode, which can effectively shorten the size of the millimeter wave transmission system. This requires the $TE_{02}-TE_{01}$ mode converter built in the tube body to have the characteristics of compact and efficient conversion.

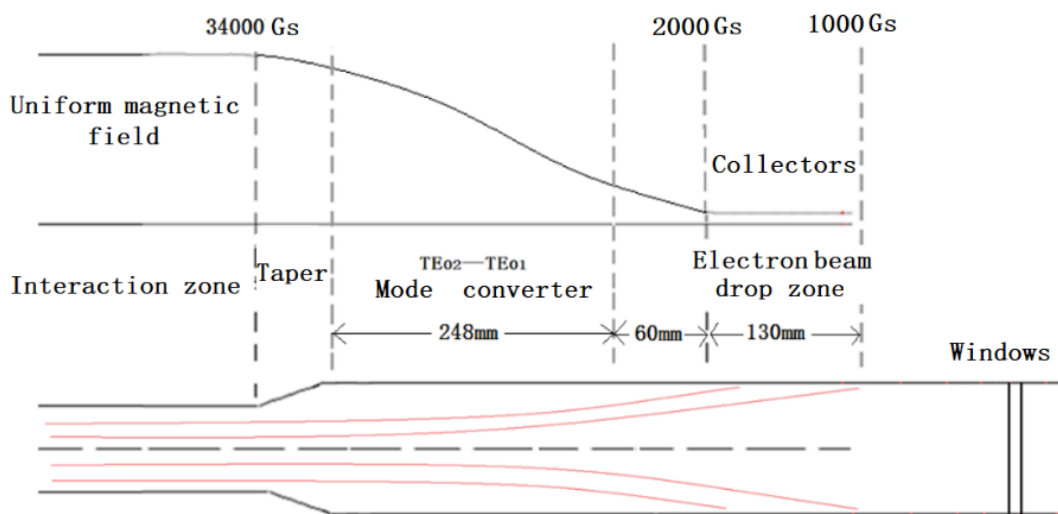


Fig. 9 collector magnetic field and electron beam trajectory diagram

In order to ensure the completion of the $TE_{02}-TE_{01}$ converter within the length of $250mm$, the waveguide radius is $9mm$.



Fig. 10 TE₀₂-TE₀₁ mode converter

The bandwidth of the mode converter is 215.8mm. Figure 10 shows the mode converter. The bandwidth is 5.9GHz when the conversion efficiency is 90%.

2) Test results

The test is carried out under the conditions of working voltage 65kV, current 10A and magnetic field 3.4 Tesla. Figure 11 is the voltage, current, input signal, and output signal shown on oscilloscope during the test. The frequency-output power curve obtained from the test in Figure 12 shows a peak output power of 120kW at a frequency of 92.4GHz and a 3dB bandwidth of 4.5GHz. In the test, when the duty is increased to the cut-off section when the electron beam bombardment occurs when TE₀₁ mode is used as the working mode, there is no electron bombardment to the cut-off section. After that, the duty continued to be increased, and the Gyro-TWT worked normally.

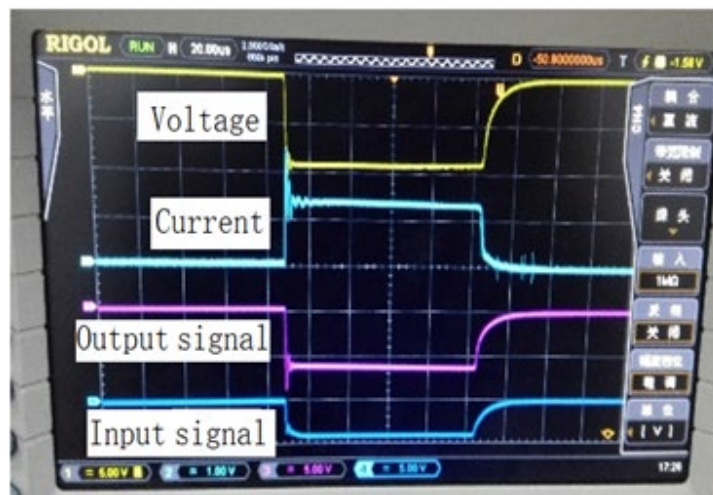


Fig. 11 Signal waveform diagram during test

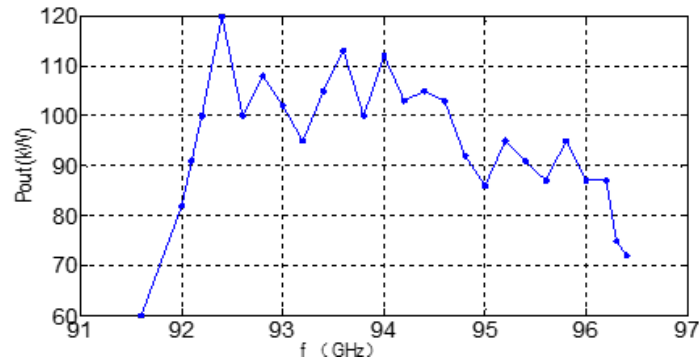


Fig. 12 Power-frequency relationship diagram during test

Comparing the simulation results with the actual test results, the test results are much smaller in terms of power and bandwidth than the simulation results. The reason is that the velocity dispersion of the electron beam is not considered in the simulation, and the velocity dispersion of the electron gun in the is designed to be 5.6%. After the actual assembly, the velocity dispersion of the electronic beam is much larger than the design value, which causes the test result to be much smaller than the simulation result.

5. Conclusions

For the W-band gyrotron traveling wave, the magnetic field is high, adiabatic compression zone of the electron gun is long, and the interaction radius is small. When TE_{01} mode is used, it is easy to cause the electron beam to bombard the cut-off section. Therefore, the TE_{02} mode is used to increase the interaction radius and improve the electron flow rate. In the experiment, the high-frequency structure adopts the periodic dielectric loading structure, which solves the problem of electron bombardment to the cut-off section, and obtains 120kW peak power, 3dB bandwidth 4.5GHz, The mode converter placed in the collector converts the working mode TE_{02} into TE_{01} output.

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